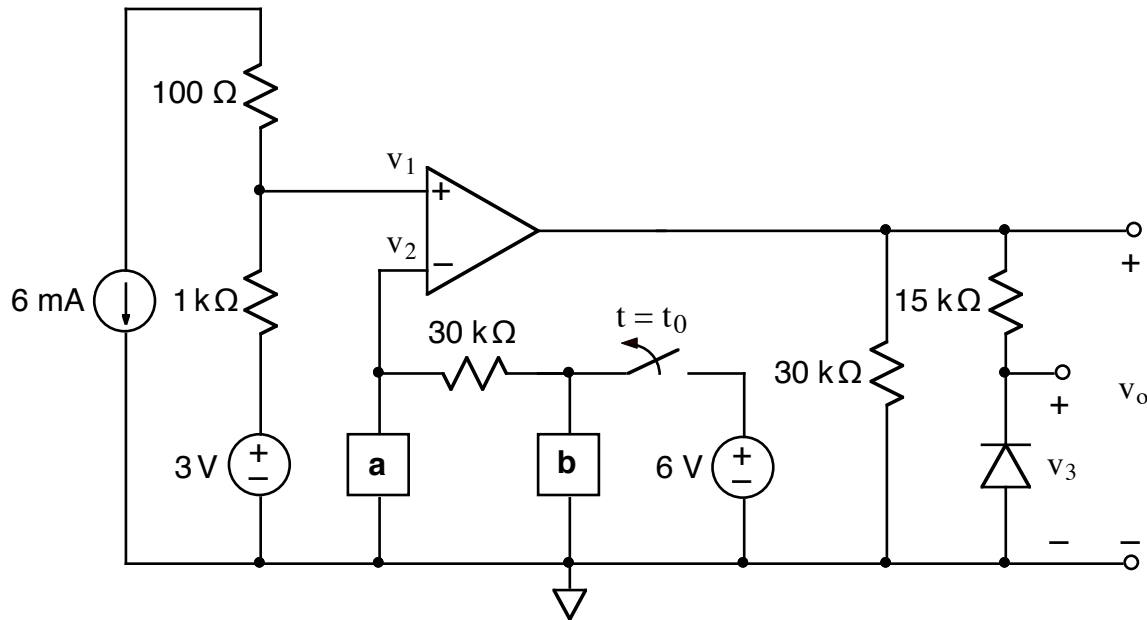
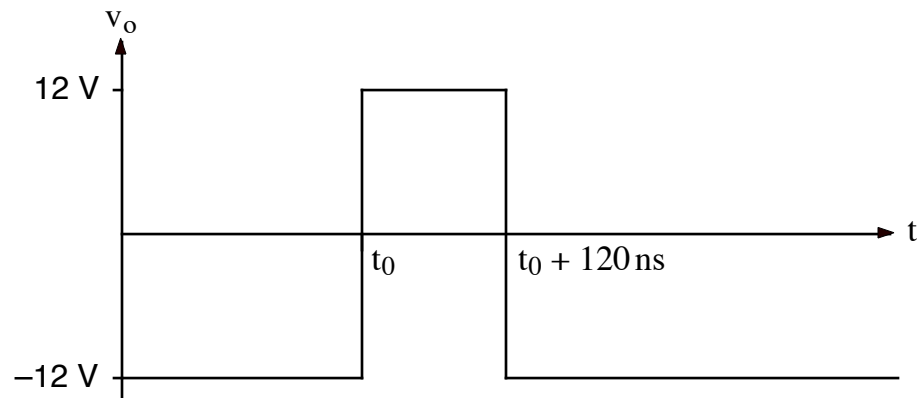


Ex:



Rail voltages = ± 12 V

After being closed for a long time, the switch opens at $t = t_0$.

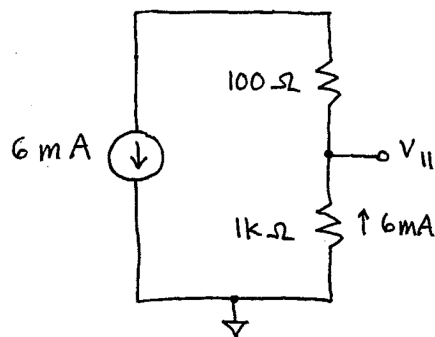


- Choose either an R or L to go in box **a** and either an R or L to go in box **b** to produce the $v_o(t)$ shown above. (Note that v_o stays low forever after $t_0 + 120$ ns.) Specify which element goes in each box and its value. Make sure -10 V $\leq v_2 \leq 10$ V at all times.
- Sketch $v_1(t)$, showing numerical values appropriately.
- Sketch $v_2(t)$, showing numerical values appropriately.
- Sketch $v_3(t)$. Show numerical values for $t < t_0$, for $t_0 < t < t_0 + 120$ ns, and for $t > t_0 + 120$ ns. Use the ideal model of the diode: when forward biased, its resistance is zero; when reverse biased, its resistance is infinite.

Soln: a) First we calculate V_1 . The op-amp input draws no current, implying that we may ignore the op-amp when calculating V_1 .

Using superposition, we sum the values of V_1 obtained when one source at a time is turned on.

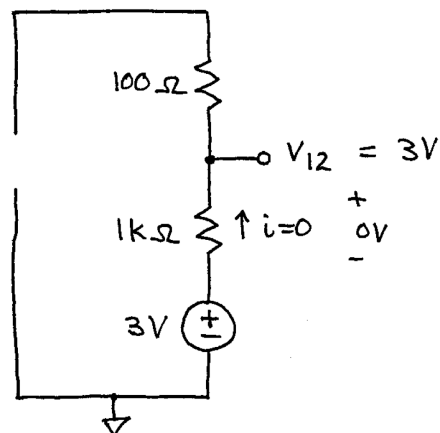
Case I: 6 mA on, 3V off



$$V_{11} = V \text{ drop across } 1k\Omega.$$

$$V_{11} = -6\text{mA} \cdot 1k\Omega = -6V$$

Case II: 6 mA off, 3V on



We sum v_{11} and v_{12} to get v_1 .

$$v_1 = v_{11} + v_{12} = -6V + 3V = -3V$$

For v_o to be $-12V$ at $t=t_0^-$ we must have $v_2 > v_1$. This condition is satisfied by placing either an L or R in box **a**. If we have an L in box **a**, then it will act like a wire at $t=t_0^-$, making v_2 equal to $0V$. If we an R in box **a**, however, we will have a v-divider:

$$v_2 = 6V \cdot \frac{R}{R+30k\Omega} > 0V$$

Now we consider what might be in box **b**. If **a** is an L and **b** is an R, then $i_L(t_0^+) = i_L(t_0^-)$ will flow thru the $30k\Omega$ and the R in **b** when the switch opens. The direction of current flow will result in $v_2 < 0V$. For sufficiently large R, we can have $v_2(t_0^+) < v_1$. Later, as the current in the L decays toward $0A$, we will again have $v_2 < v_1$, causing v_o to drop to $-12V$ again. Thus, this sol'n will work.

$$\mathbf{a} = \mathbf{L} \quad \mathbf{b} = \mathbf{R}$$

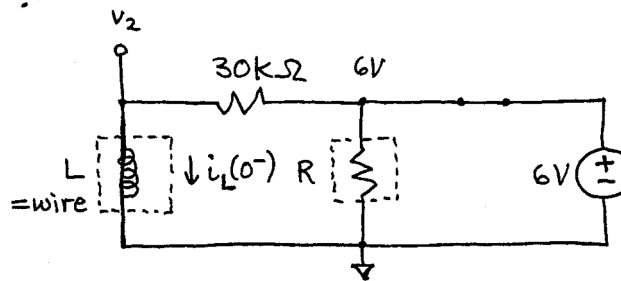
To be thorough, we consider what would happen for the case **a** = R and **b** = L. This would cause a short across the $6V$ source for $t < t_0$ since $L = \text{wire}$. This will not work!

Now we find $v_2(t)$ using the general form of solution for RL problems:

$$v_2(t > 0) = v_2(t \rightarrow \infty) + [v_2(0^+) - v_2(t \rightarrow \infty)] e^{-t / \frac{L}{R_{TH}}}$$

where we take t_0 to be $t_0 = 0$ s.

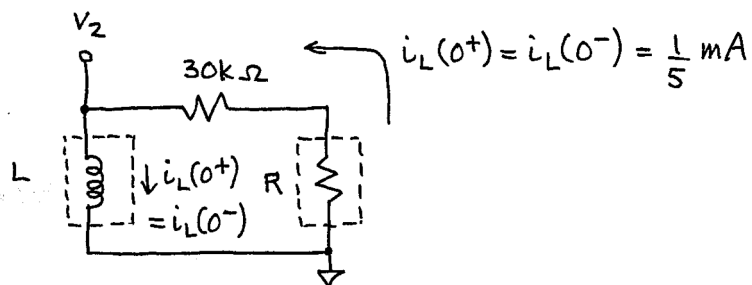
$t = 0^-$:



The R is a 2nd circuit, (in parallel with the L and $30k\Omega$), across the 6V source. Thus, we may ignore R.

$$i_L(0^-) = \frac{6V}{30k\Omega} = \frac{1}{5} \text{ mA}$$

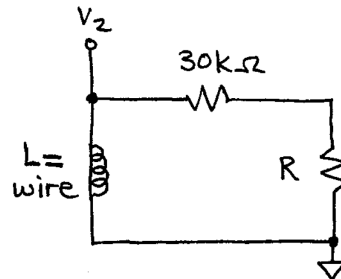
$t = 0^+$:



$$v_2(0^+) = -i_L(0^+) (30k\Omega + R)$$

$$\text{or } v_2(0^+) = -\frac{1}{5} \text{ mA} (30k\Omega + R)$$

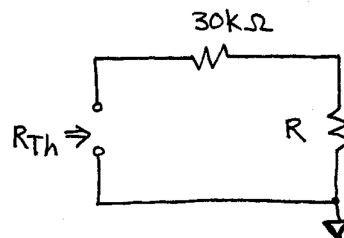
$t \rightarrow \infty$:



$v_2(t \rightarrow \infty) = 0V$ since there is no power source

R_{Th} for $\frac{L}{R_{Th}}$ time constant = R seen

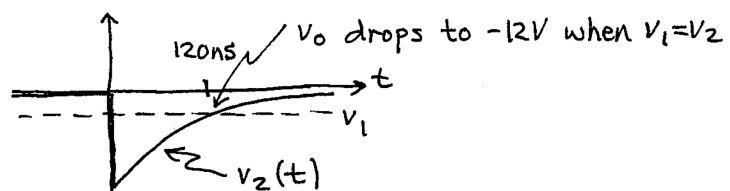
looking into terminals where L is attached, (but with L removed):



$$R_{Th} = 30k\Omega + R$$

$$\therefore v_2(t > 0) = 0V + \left[-\frac{1}{5}mA(30k + R) - 0V \right] e^{-\frac{t}{L/(30k\Omega + R)}}$$

sketch:



We choose L and R values that yield $v_2(t) = v_1$ at $t = 120 \text{ ns}$, the duration of the pulse. Recall that $v_1 = -3 \text{ V}$.

$$v_2(t=120 \text{ ns}) = -\frac{1 \text{ mA}}{5}(30 \text{ k}\Omega + R) e^{-\frac{t=120 \text{ ns}}{L/(30 \text{ k}\Omega + R)}} = -3 \text{ V}$$

We cannot solve this eq'n uniquely for R and L. Since it is difficult to solve for R, we choose a value of R and then solve for L.

We must have $v_2(0^+) < -3 \text{ V}$ to get a pulse for v_o .

$$\therefore -\frac{1}{5} \text{ mA}(30 \text{ k}\Omega + R) < -3 \text{ V}$$

We see that any R will, ^{work} except $R=0$ that would short out the 6 V source for $t < t_0$.

A convenient value for R is $20 \text{ k}\Omega$. This gives

$$v_2(0^+) = -\frac{1}{5} \text{ mA}(30 \text{ k}\Omega + 20 \text{ k}\Omega) = -10 \text{ V}$$

Now we can solve for L and see if we get a reasonable value.

$$v_2(t=120 \text{ ns}) = -10 \text{ V} e^{-\frac{120 \text{ ns}}{L/50 \text{ k}\Omega}} = -3 \text{ V}$$

$$\text{or } e^{-\frac{120 \text{ ns}}{L/50 \text{ k}\Omega}} = \frac{-3 \text{ V}}{-10 \text{ V}} = \frac{3}{10}$$

$$\text{or } \frac{-120 \text{ ns}}{L/50 \text{ k}\Omega} = \ln\left(\frac{3}{10}\right)$$

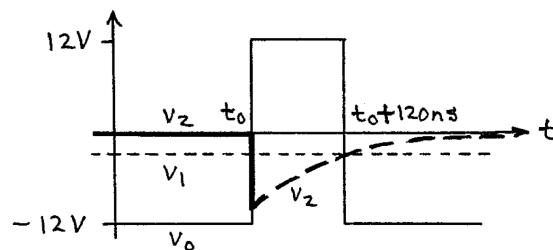
$$\text{or } \frac{L}{50k\Omega} = \frac{-120\text{ns}}{\ln\left(\frac{3}{10}\right)} = \frac{-120\text{ns}}{-1.204} = 99.7\text{ns}$$

$$\text{or } \frac{L}{50k\Omega} \approx 100\text{ns}$$

$$\text{or } L \approx 100\text{n} \cdot 50k\Omega = 5\text{mH}$$

Summary: $R = 20k\Omega$, $L = 5\text{mH}$

b) and c) We use the above results to make our plot:



d) The diode acts like a wire when current flows in the direction of the arrow in the diode symbol. Otherwise, the diode is an open circuit.

For $t < t_0$, $v_0 = -12\text{V}$ so the diode looks like a wire shorting v_3 to 0V. Applies to $t > t_0 + 120\text{ns}$, too.
For $t > t_0$ & $v_0 = +12\text{V}$, the diode acts like an open, and v_3 is pulled up to $v_0 = 12\text{V}$ by the $15k\Omega$ resistor, which has no V-drop.

